

# Direction-of-Arrival Estimation for Correlated Sources and Low Sample Size

**ETH** zürich

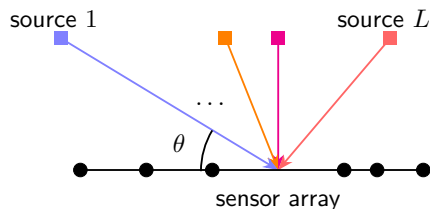


TECHNISCHE  
UNIVERSITÄT  
DARMSTADT

Yani Zhang,  
ETH Zurich,  
Tianyi Liu, Marius Pesavento,  
TU Darmstadt

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# Direction-of-Arrival (DOA) estimation



- Classic problem in e.g. wireless communication, seismic exploration, automatic monitoring
- Numerous approaches:  
maximum likelihood estimator [Stoica, 1989], root-MUSIC [Rao, 1989], ESPRIT [Paulraj, Roy, Kailath, 1989], ...
- **Very difficult: correlated sources and low sample size**
  - Partial Relaxation (PR) [Trinh-Hoang, 2018]
  - SPARse ROW-norm reconstruction (SPARROW) [Steffens, 2018]

# Signal model

- $M$  omnidirectional sensors, narrowband signals located in the farfield of the array
- $L$  source signals with DOAs  $\boldsymbol{\theta} = [\theta_1, \dots, \theta_L]^\top$  ( $L$  is known)
- Received signals

$$\mathbf{y}(t) = \mathbf{A}(\boldsymbol{\theta})\mathbf{x}(t) + \mathbf{n}(t)$$

- $\mathbf{y}(t) \in \mathbb{C}^M$  received signal vector
- $\mathbf{x}(t) \in \mathbb{C}^L$  source signal vector
- $\mathbf{A}(\boldsymbol{\theta}) = [\mathbf{a}(\theta_1), \dots, \mathbf{a}(\theta_L)] \in \mathbb{C}^{M \times L}$  steering matrix
- $\mathbf{a}(\theta_i) = [1, e^{-j\pi \sin(\theta_i)}, \dots, e^{-j(M-1)\pi \sin(\theta_i)}]^\top$  sensor array responses for DOA  $\theta_i$  with uniform linear array
- $\mathbf{n}(t) \in \mathbb{C}^M$

- $M$  omnidirectional sensors, narrowband signals located in the farfield of the array
- $L$  source signals with DOAs  $\boldsymbol{\theta} = [\theta_1, \dots, \theta_L]^\top$  ( $L$  is known)
- Multiple snapshots

$$\mathbf{Y} = \mathbf{A}(\boldsymbol{\theta})\mathbf{X} + \mathbf{N}$$

- $\mathbf{Y} = [\mathbf{y}(1), \dots, \mathbf{y}(N)]$ : received signal matrix
- $\mathbf{X} = [\mathbf{x}(1), \dots, \mathbf{x}(N)]$ : source signal matrix
- $\mathbf{N} = [\mathbf{n}(1), \dots, \mathbf{n}(N)]$ : noise matrix

# Signal model

- Source signals: stationary process
  - $\mathbb{E}\{\mathbf{x}(t)\} = \mathbf{0}$
  - source covariance matrix  $\mathbf{R}_x = \mathbb{E}\{\mathbf{x}(t)\mathbf{x}(t)^H\}$
- Noise: i.i.d. white Gaussian
  - $\mathbf{R}_n = \mathbb{E}\{\mathbf{n}(t)\mathbf{n}(t)^H\} = \sigma^2 \mathbf{I}_M$
  - $\sigma^2$ : noise power at each sensor
- Received signals
  - $\mathbf{R}_y = \mathbb{E}\{\mathbf{y}(t)\mathbf{y}(t)^H\} = \mathbf{A}(\boldsymbol{\theta})\mathbf{R}_x\mathbf{A}(\boldsymbol{\theta})^H + \sigma^2 \mathbf{I}_M$
  - $\widehat{\mathbf{R}}_y = \frac{1}{N} \mathbf{Y}\mathbf{Y}^H$

# Partial Relaxation

- Conventional

$$\hat{\theta} = \arg \min_{\theta} f(\mathbf{A}(\theta), \mathbf{Y})$$



- PR [Trinh-Hoang, 2018]

$$\hat{\theta} = {}^L \arg \min_{\theta} \min_{\mathbf{B} \in \mathbb{C}^{M \times (L-1)}} f(\mathbf{a}(\theta), \mathbf{B}, \mathbf{Y})$$

where  ${}^L \arg \min$  : take  $L$  deepest minima

- degrade severely for highly correlated sources or low sample size



- Sparse signal reconstruction

$$\min_{\mathbf{Z} \in \mathbb{C}^{K \times N}} \frac{1}{2} \|\mathbf{A}(\boldsymbol{\nu})\mathbf{Z} - \mathbf{Y}\|_{\mathbb{F}}^2 + \lambda \sqrt{N} \|\mathbf{Z}\|_{2,1} \quad (1)$$

- $\boldsymbol{\nu} = \{\nu_1, \dots, \nu_K\}$ : sampled FOV with  $K \gg L$
- $\mathbf{Z} = [\mathbf{z}_1, \dots, \mathbf{z}_K]^T$ : row-wise sparse signal matrix
- $\mathbf{A}(\boldsymbol{\nu}) = [\mathbf{a}(\nu_1), \dots, \mathbf{a}(\nu_K)] \in \mathbb{C}^{M \times K}$ : overcomplete dictionary matrix
- $\ell_{2,1}$ -mixed-norm

$$\|\mathbf{Z}\|_{2,1} = \sum_{k=1}^K \|\mathbf{z}_k\|_2$$

- $\lambda > 0$ : regularization parameter inducing row-sparsity in  $\mathbf{Z}$
- On-grid assumption:  $\theta_i \in \boldsymbol{\nu}$ , for  $i = 1, \dots, L$
- Jointly estimating the sources: robust to correlated sources

- Sparse signal reconstruction

$$\min_{\mathbf{Z} \in \mathbb{C}^{K \times N}} \frac{1}{2} \|\mathbf{A}(\boldsymbol{\nu})\mathbf{Z} - \mathbf{Y}\|_{\text{F}}^2 + \lambda\sqrt{N}\|\mathbf{Z}\|_{2,1} \quad (1)$$

- Problem (1) is equivalent to the convex problem

$$\min_{\mathbf{S} \in \mathbb{D}_+^K} \text{Tr} \left( (\mathbf{A}(\boldsymbol{\nu})\mathbf{S}\mathbf{A}(\boldsymbol{\nu})^{\text{H}} + \lambda\mathbf{I}_M)^{-1} \widehat{\mathbf{R}}_y \right) + \text{Tr}(\mathbf{S}) \quad (2)$$

- $\mathbb{D}_+^K$ : the set of  $K \times K$  nonnegative diagonal matrices.

- The solutions of (1) and (2) are related by

$$\widehat{\mathbf{Z}} = \widehat{\mathbf{S}}\mathbf{A}(\boldsymbol{\nu})^{\text{H}}(\mathbf{A}(\boldsymbol{\nu})\widehat{\mathbf{S}}\mathbf{A}(\boldsymbol{\nu})^{\text{H}} + \lambda\mathbf{I}_M)^{-1}\mathbf{Y},$$

$$\widehat{s}_k = \frac{1}{\sqrt{N}} \|\widehat{\mathbf{z}}_k\|_2 \quad \text{for } k = 1, 2, \dots, K,$$

with  $s_1, \dots, s_K \geq 0$  being the diagonal entries of  $\mathbf{S}$ .



# Gridless SPARROW

- Uniform linear array with  $M$  sensors
  - $\mathbf{A}(\boldsymbol{\nu})$  has a Vandemonde structure
  - $\mathbf{S}$  nonnegative diagonal
  - positive semidefinite Toeplitz

$$\mathbf{T} = \mathbf{A}(\boldsymbol{\nu})\mathbf{S}\mathbf{A}(\boldsymbol{\nu})^H = \sum_{k=1}^K s_k \mathbf{a}(\nu_k)\mathbf{a}(\nu_k)^H \in \mathcal{T}_+^M$$

- $\mathcal{T}_+^M$  : the set of  $M \times M$  positive semidefinite Toeplitz matrices
- Gridless SPARROW [Steffens, 2018]

$$\hat{\mathbf{T}} = \arg \min_{\mathbf{T} \in \mathcal{T}_+^M} \text{Tr} \left( (\mathbf{T} + \lambda \mathbf{I}_M)^{-1} \hat{\mathbf{R}}_y \right) + \frac{1}{M} \text{Tr}(\mathbf{T}) \quad (3)$$

- Vandemonde decomposition on  $\hat{\mathbf{T}}$  to recover DOAs

# Proposed method

- Step 1: solve gridless SPARROW with a **reduced regularization parameter**

$$\hat{\mathbf{T}} = \arg \min_{\mathbf{T} \in \mathcal{T}_+^M} \text{Tr} \left( (\mathbf{T} + \lambda \mathbf{I}_M)^{-1} \hat{\mathbf{R}}_y \right) + \frac{1}{M} \text{Tr}(\mathbf{T})$$

- $\lambda = C_\lambda \underbrace{\sqrt{\sigma^2 M \log M}}_{\text{empirical}}, \quad C_\lambda = 0.4 \sim 0.6$
- **reduced bias**

- Step 2: use **PR** on  $\hat{\mathbf{T}}$  to recover DOAs
  - extract signal subspace from  $\hat{\mathbf{T}}$

# Simulation results

- Setup:

- Signal-to-noise ratio:  $\text{SNR}=1/\sigma^2$
- Root-mean-square-error:

$$\text{RMSE} = \sqrt{\frac{1}{N_R L} \sum_{i=1}^{N_R} \sum_{l=1}^L (\hat{\theta}_l^{(i)} - \theta_l)^2},$$

- Monte-Carlo runs: 200

- Estimators:

- The proposed method
- root-MUSIC
- PR
- SPARROW with the original regularization ( $C_\lambda = 1$ )
- SPARROW with the same regularization as the proposed method ( $C_\lambda = 0.4$ )
- root-MUSIC applied to the Toeplitz matrix solution with regularization  $C_\lambda = 0.4$

# Simulation 1: few snapshots

- Two **uncorrelated** sources at  $-30^\circ$  and  $30^\circ$
- Snapshots  $N = 3$ , uniform linear array of size  $M = 6$

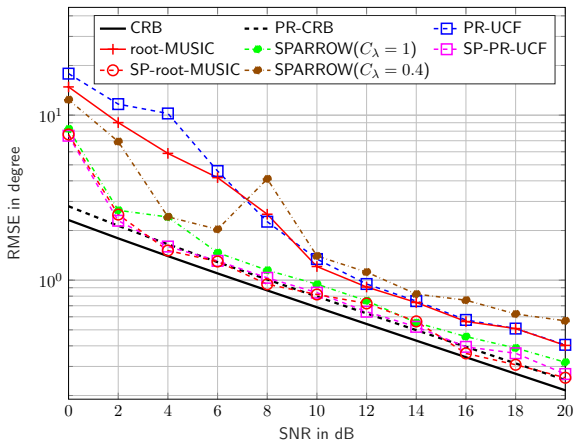


Fig. 1. RMSE vs SNR

## Simulation 2: highly-correlated source signals

- Two sources at  $45^\circ$  and  $50^\circ$ , correlation factor  $\rho = 0.95$
- Uniform linear array of size  $M = 10$

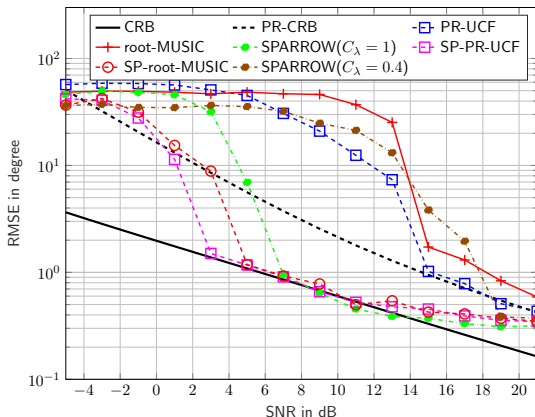


Fig. 2. RMSE vs SNR with snapshots  $N = 40$

## Simulation 2: highly-correlated source signals

- Two sources at  $45^\circ$  and  $50^\circ$ , correlation factor  $\rho = 0.95$
- Uniform linear array of size  $M = 10$

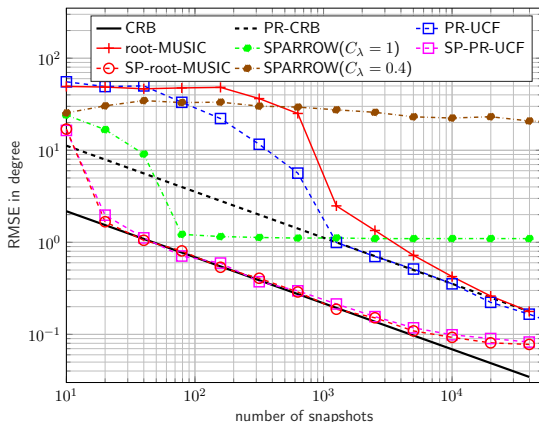


Fig. 3. RMSE vs  $N$  with SNR= 5 dB

## Simulation 2: highly-correlated source signals

- Two sources at  $45^\circ$  and  $50^\circ$ , correlation factor  $\rho = 0.95$
- Uniform linear array of size  $M = 10$

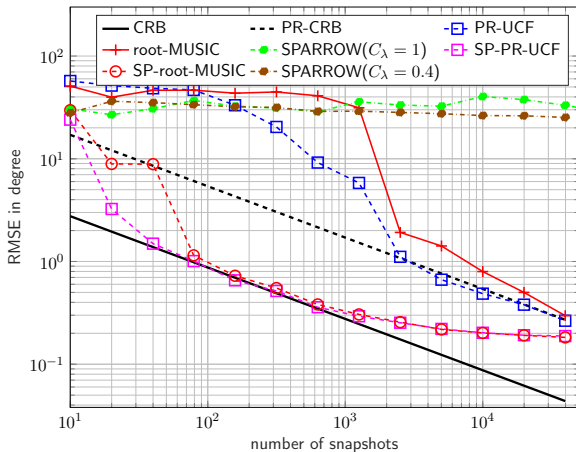


Fig. 4. RMSE vs  $N$  with SNR= 3 dB

# Conclusion

- Take away:
  - Step 1 gridless SPARROW, Step 2 PR
  - robust w.r.t. highly correlated sources and low sample size, superior than both PR and SPARROW
- Generalizes from uniform linear arrays to other array geometries